

HGTG5N120CND, HGTP5N120CND, HGT1S5N120CNDS

Data Sheet

December 2001

25A, 1200V, NPT Series N-Channel IGBT with Anti-Parallel Hyperfast Diode

The HGTG5N120CND, HGTP5N120CND and HGT1S5N120CNDS are **N**on-**P**unch **T**hrough (NPT) IGBT designs. They are new members of the MOS gated high voltage switching IGBT family. IGBTs combine the best features of MOSFETs and bipolar transistors. This device has the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The IGBT used is developmental type TA49309. The diode used in anti-parallel is developmental typeTA49058.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

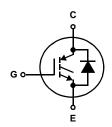
Formerly Developmental Type TA49307.

Ordering Information

PART NUMBER	PACKAGE	BRAND
HGTG5N120CND	TO-247	5N120CND
HGTP5N120CND	TO-220AB	5N120CND
HGT1S5N120CNDS	TO-263AB	5N120CND

NOTE: When ordering, use the entire part number. Add the suffix 9A to obtain the TO-263AB variant in Tape and Reel, i.e., HGT1S5N120CNDS9A.

Symbol



Features

- 25A, 1200V, T_C = 25^oC
- 1200V Switching SOA Capability
- · Short Circuit Rating
- Low Conduction Loss
- Temperature Compensating SABER™ Model Thermal Impedance SPICE Model www.fairchildsemi.com
- · Related Literature
 - TB334 "Guidelines for Soldering Surface Mount Components to PC Boards"

Packaging

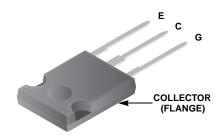
JEDEC TO-220AB ALTERNATE VERSION



JEDEC TO-263AB



JEDEC STYLE TO-247



FAIRCH	IILD SEMICONDU	ICTOR IGBT PRO	DUCT IS COVERI	ED BY ONE OR M	ORE OF THE FO	LLOWING U.S. PA	ATENTS
4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,587,713
4,598,461	4,605,948	4,620,211	4,631,564	4,639,754	4,639,762	4,641,162	4,644,637
4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690	4,794,432	4,801,986
4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606	4,860,080	4,883,767
4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951	4,969,027	

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Absolute Maximum Ratings $T_C = 25^{\circ}C$, Unless Otherwise Specified

HGTG5N120CND HGTP5N120CND HGT1S5N120CNDS	UNITS
1200	V
25	Α
12	Α
40	Α
±20	V
±30	V
30A at 1200V	
167	W
1.33	W/oC
-55 to 150	οС
300	οС
260	οС
8	μs
15	μs
	HGTP5N120CND HGT1S5N120CNDS 1200 25 12 40 ±20 ±30 30A at 1200V 167 1.33 -55 to 150 300 260 8

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTES

- 1. Pulse width limited by maximum junction temperature.
- 2. $V_{CE(PK)} = 840V$, $T_J = 125^{o}C$, $R_G = 25\Omega$.

Electrical Specifications $T_C = 25^{\circ}C$, Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNITS
Collector to Emitter Breakdown Voltage	BV _{CES}	$I_C = 250 \mu A, V_{GE} = 0 V$		1200	-	-	V
Collector to Emitter Leakage Current	I _{CES}	V _{CE} = 1200V	$T_C = 25^{\circ}C$	-	-	250	μΑ
			$T_{C} = 125^{\circ}C$	-	100	-	μΑ
			$T_{\rm C} = 150^{\rm o}{\rm C}$	-	-	2	mA
Collector to Emitter Saturation Voltage	V _{CE(SAT)}	$I_C = 5.5A,$ $T_C = 25^{\circ}$	$T_{\rm C} = 25^{\rm o}{\rm C}$	-	2.1	2.4	V
		$V_{GE} = 15V$	$T_{C} = 150^{\circ}C$	-	2.9	3.5	V
Gate to Emitter Threshold Voltage	V _{GE(TH)}	$I_C = 45\mu A$, $V_{CE} = V_{GE}$		6.0	7.0	-	V
Gate to Emitter Leakage Current	I _{GES}	V _{GE} = ±20V		-	-	±250	nA
Switching SOA	SSOA	$T_J = 150^{\circ}\text{C}, R_G = 25\Omega, V_{GE} = 15V, L = 200\mu\text{H}, V_{CE(PK)} = 1200V$		25	-	-	А
Gate to Emitter Plateau Voltage	V _{GEP}	I _C = 5.5A, V _{CE} = 600V		-	10.6	-	V
On-State Gate Charge	Q _{G(ON)}	$I_C = 5.5A,$ $V_{GE} = 15V$		-	45	55	nC
		V _{CE} = 600V	V _{GE} = 20V	-	60	75	nC
Current Turn-On Delay Time	t _d (ON)I	IGBT and Diode at T _J = 25°C		-	22	30	ns
Current Rise Time	t _{rl}	$R_G = 25\Omega$ $L = 5mH$ Test Circuit (Figure 20)			16	ns	
Current Turn-Off Delay Time	t _{d(OFF)I}				250	ns	
Current Fall Time	t _{fl}				350	ns	
Turn-On Energy	E _{ON}				500	μJ	
Turn-Off Energy (Note 3)	E _{OFF}		-	640	700	μJ	

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Electrical Specifications $T_C = 25^{\circ}C$, Unless Otherwise Specified (Continued)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Current Turn-On Delay Time	^t d(ON)I	IGBT and Diode at T _J = 150 ^o C	-	20	25	ns
Current Rise Time	t _{rl}	$I_{CE} = 5.5A$ $V_{CF} = 960V$	-	12	16	ns
Current Turn-Off Delay Time	t _{d(OFF)I}	$V_{GE} = 15V$ $R_{G} = 25\Omega$	-	225	300	ns
Current Fall Time	t _{fl}	L = 5mH	-	350	400	ns
Turn-On Energy	E _{ON}	Test Circuit (Figure 20)		1	1.2	mJ
Turn-Off Energy (Note 3)	E _{OFF}		-	1	1.1	mJ
Diode Forward Voltage	V _{EC}	I _{EC} = 5.5A	-	2.4	3.3	V
Diode Reverse Recovery Time	t _{rr}	$I_{EC} = 5.5A$, $dI_{EC}/dt = 200A/\mu s$	-	48	60	ns
		$I_{EC} = 1A$, $dI_{EC}/dt = 200A/\mu s$	-	30	40	ns
Thermal Resistance Junction To Case	$R_{ heta JC}$	IGBT	-	-	0.75	oC/M
		Diode	-	-	1.9	oC/W

NOTE:

Typical Performance Curves Unless Otherwise Specified

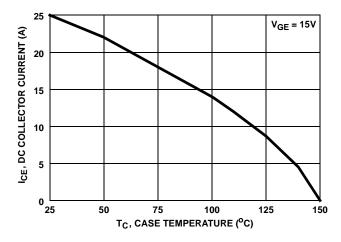


FIGURE 1. DC COLLECTOR CURRENT vs CASE TEMPERATURE

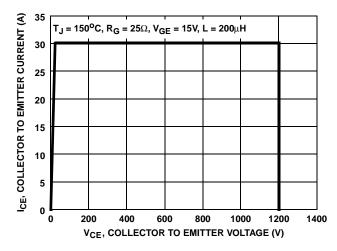


FIGURE 2. MINIMUM SWITCHING SAFE OPERATING AREA

^{3.} Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I_{CE} = 0A). All devices were tested per JEDEC Standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

Typical Performance Curves Unless Otherwise Specified (Continued)

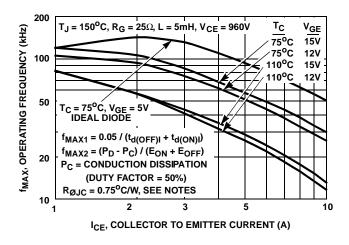


FIGURE 3. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

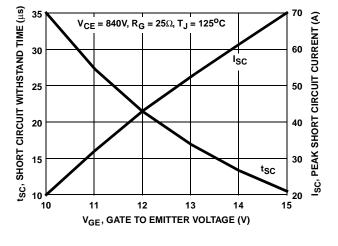


FIGURE 4. SHORT CIRCUIT WITHSTAND TIME

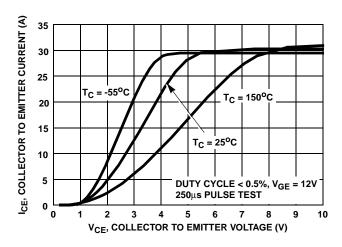


FIGURE 5. COLLECTOR TO EMITTER ON-STATE VOLTAGE

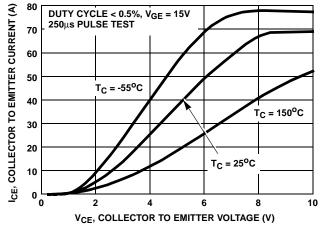


FIGURE 6. COLLECTOR TO EMITTER ON-STATE VOLTAGE

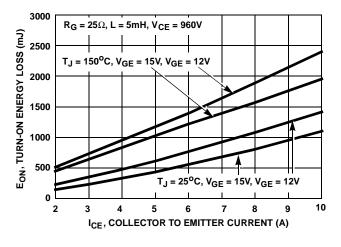


FIGURE 7. TURN-ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

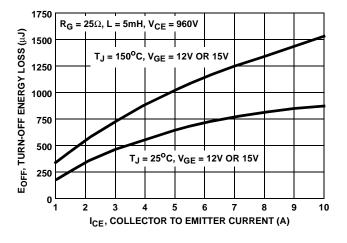


FIGURE 8. TURN-OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

Typical Performance Curves Unless Otherwise Specified (Continued)

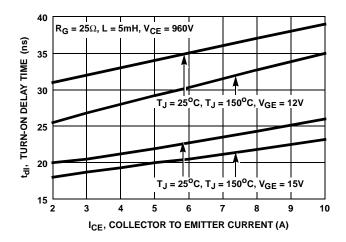


FIGURE 9. TURN-ON DELAY TIME vs COLLECTOR TO EMITTER CURRENT

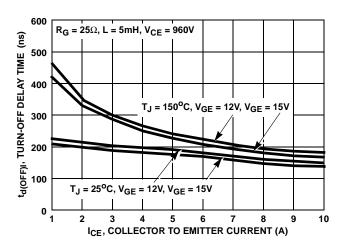


FIGURE 11. TURN-OFF DELAY TIME vs COLLECTOR TO EMITTER CURRENT

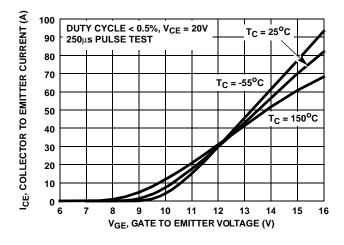


FIGURE 13. TRANSFER CHARACTERISTIC

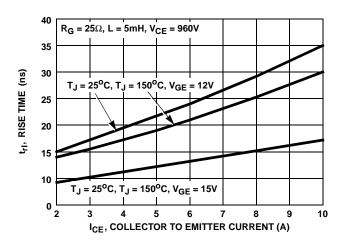


FIGURE 10. TURN-ON RISE TIME vs COLLECTOR TO EMITTER CURRENT

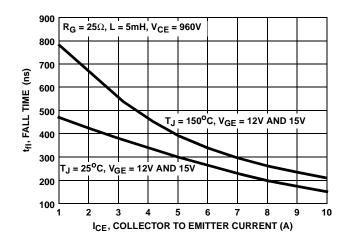


FIGURE 12. FALL TIME vs COLLECTOR TO EMITTER CURRENT

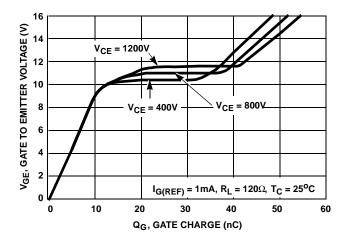


FIGURE 14. GATE CHARGE WAVEFORMS

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Typical Performance Curves Unless Otherwise Specified (Continued)

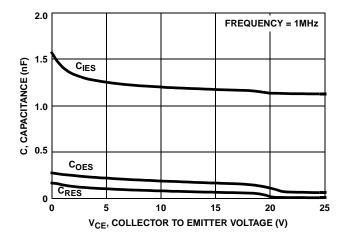


FIGURE 15. CAPACITANCE vs COLLECTOR TO EMITTER VOLTAGE

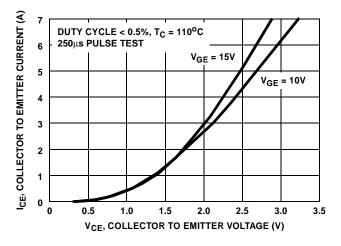


FIGURE 16. COLLECTOR TO EMITTER ON-STATE VOLTAGE

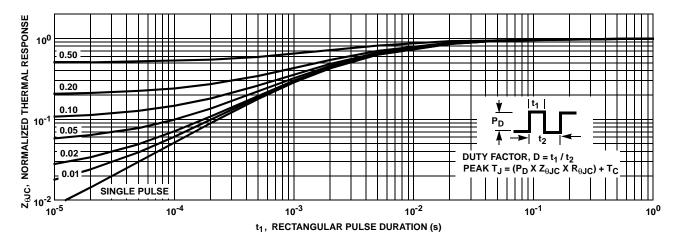


FIGURE 17. NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

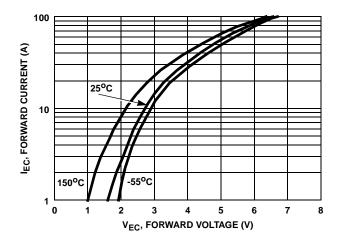


FIGURE 18. DIODE FORWARD CURRENT vs FORWARD VOLTAGE DROP

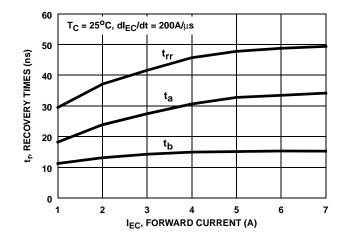


FIGURE 19. RECOVERY TIMES vs FORWARD CURRENT

Test Circuit and Waveforms

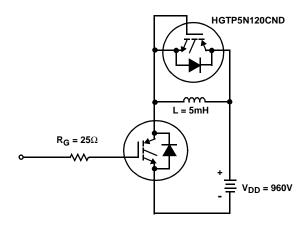


FIGURE 20. INDUCTIVE SWITCHING TEST CIRCUIT

Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- Devices should never be inserted into or removed from circuits with power on.
- Gate Voltage Rating Never exceed the gate-voltage rating of V_{GEM}. Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.
- 6. **Gate Termination** The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- Gate Protection These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

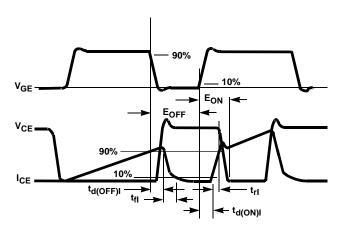


FIGURE 21. SWITCHING TEST WAVEFORMS

Operating Frequency Information

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows f_{MAX1} or f_{MAX2} ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 f_{MAX1} is defined by $f_{MAX1}=0.05/(t_{d(OFF)I}+t_{d(ON)I}).$ Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{d(OFF)I}$ and $t_{d(ON)I}$ are defined in Figure 21. Device turn-off delay can establish an additional frequency limiting condition for an application other than $T_{JM},\ t_{d(OFF)I}$ is important when controlling output ripple under a lightly loaded condition.

 f_{MAX2} is defined by f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON}). The allowable dissipation (P_D) is defined by P_D = (T_{JM} - T_C)/R_{\theta JC}. The sum of device switching and conduction losses must not exceed P_D . A 50% duty factor was used (Figure 3) and the conduction losses (P_C) are approximated by P_C = (V_CE x I_CE)/2.

 E_{ON} and E_{OFF} are defined in the switching waveforms shown in Figure 21. E_{ON} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-off. All tail losses are included in the calculation for E_{OFF} ; i.e., the collector current equals zero ($I_{CE} = 0$).

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